

## RELATED PATENT APPLICATIONS

This application is related to U.S. Pate	nt
Application Serial No entitled "Meth	ıod
and System for Transmitting Information in an Option	al
Communication System Using Distributed Amplification	ı, <sub>ji</sub>
U.S. Patent Application Serial No.	
entitled "Receiver and Method for a Multichannel Option	al
Communication System, " U.S. Patent Application Serial N	10.
entitled "Method and System f	or
Demultiplexing Non-Intensity Modulated Waveleng	jth
Division Multiplexed (WDM) Signals," and U.S. Pate	nt
Application Serial No entitled "Meth	ıod
and System for Tuning an Optical Signal Based	on
Transmission Conditions, " all filed on, 2001.	

## TECHNICAL FIELD OF THE INVENTION

The present invention relates generally to optical communication systems, and more particularly to a method and system for communicating a clock signal over a optical link.

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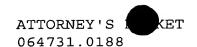
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#### BACKGROUND OF THE INVENTION

Telecommunications systems, cable television systems and data communication networks use optical networks to rapidly convey large amounts of information between remote points. In an optical network, information is conveyed in the form of optical signals through optical fibers. Optical fibers are thin strands of glass capable of transmitting the signals over long distances with very low loss.

Optical networks often employ wavelength division multiplexing (WDM) to increase transmission capacity. In a WDM network, a number of optical channels are carried in each fiber at disparate wavelengths. Network capacity is increased as a multiple of the number of wavelengths, or channels, in each fiber.

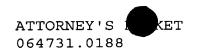
Each optical channel carries a data signal and a corresponding clock signal. The clock signal recovered and used to make forward error corrections (FEC) decisions for the data signal. FEC typically requires good clock extraction which is difficult with very poor waveforms and/or a large amount of jitter present in conventional intensity shift keying systems. To improve clock recovery, a pure clock signal has been transmitted to each node as a data signal. This however is not cost-effective and is complicated from a network operation standpoint.

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# SUMMARY OF THE INVENTION

The present invention provides a method and system for communicating a clock signal over an optical link that substantially reduces or eliminated problems and disadvantages associated with previous systems and methods. In a particular embodiment, the clock signal is modulated onto a intensity or other suitable characteristic of a data channel and extracted from that characteristic by a receiver.

In accordance with one embodiment of the present invention, a method and system for communicating a clock link includes receiving over an optical multimodulated optical information signal including nonintensity modulation for a data signal and intensity modulation for a clock signal. The clock signal recovered based on the intensity modulation of the The nonmultimodulated optical information signal. intensity modulation for the data signal is converted to intensity modulation for the data signal. The data signal is recovered from the intensity modulation for the data signal using the clock signal.

More specifically, in accordance with a particular embodiment of the present invention, the data signal may be phase shift keyed in the multimodulated optical information signal. In this embodiment, the clock signal may be intensity shift keyed in the multimodulated optical information signal. The clock signal may be used for forward error correction (FEC) in addition to data signal recovery.

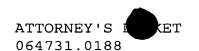
30 Technical advantages of the present invention include providing an improved method and system for communicating a clock signal over an optical link. In a particular embodiment, the clock signal is modulated onto

the intensity or other suitable characteristic of a data channel and from there extracted by a receiver. Accordingly, stability of clock recovery is improved even in the presence of poor optical signal quality and jitter. As a result, for error correction (FEC) is enhanced to lower bit error rates at the receiver.

Other technical advantages of the present invention will be readily apparent to one skilled in the art from the following figures, description and claims.

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## BRIEF DESCRIPTION OF THE DRAWINGS

For a more complete understanding of the present invention and its advantages, reference is now made to the following description taken in conjunction with the accompanying drawings, wherein like numerals represent like parts, in which:

FIGURE 1 is a block diagram illustrating an optical communication system using distributed amplification in accordance with one embodiment of the present invention;

FIGURE 2 is a block diagram illustrating the optical sender of FIGURE 1 in accordance with one embodiment of the present invention;

FIGURES 3A-C are diagrams illustrating non-intensity modulated signals for transmission in the optical communication system of FIGURE 1 in accordance with several embodiments of the present invention;

FIGURE 4 is a block diagram illustrating the optical sender of FIGURE 1 in accordance with another embodiment of the present invention;

FIGURE 5 is a diagram illustrating the optical waveform generated by the optical sender of FIGURE 4 in accordance with one embodiment of the present invention;

FIGURE 6 is a block diagram illustrating the optical receiver of FIGURE 1 in accordance with one embodiment of the present invention;

FIGURE 7 is a diagram illustrating the frequency response of the asymmetric Mach-Zender interferometer of FIGURE 6 in accordance with one embodiment of the present invention;

30 FIGURES 8A-C are block diagrams illustrating the demultiplexer of FIGURE 1 in accordance with several embodiments of the present invention;

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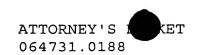


FIGURE 9 is a flow diagram illustrating a method for communicating data over an optical communication system using distributed amplification in accordance with one embodiment of the present invention;

FIGURE 10 is a block diagram illustrating a bidirectional optical communication system using distributed amplification in accordance with one embodiment of the present invention;

FIGURE 11 is a block diagram illustrating the optical sender and receiver of FIGURE 1 in accordance with another embodiment of the present invention;

FIGURE 12 is a block diagram illustrating the modulator of FIGURE 11 in accordance with one embodiment of the present invention;

FIGURE 13 is a flow diagram illustrating a method for tuning the modulation depth of an optical signal based on receiver side information in accordance with one embodiment of the present invention;

FIGURE 14 is a block diagram illustrating an optical communication system distributing a clock signal in an information channel in accordance with one embodiment of the present invention; and

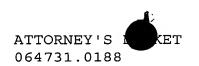
FIGURE 15 is a block diagram illustrating an optical receiver for extracting a clock signal from a multimodulated signal in accordance with one embodiment of the present invention.

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# DETAILED DESCRIPTION OF THE INVENTION

FIGURE 1 illustrates an optical communication system 10 in accordance with one embodiment of the present invention. In this embodiment, the optical communication system 10 is a wavelength division multiplexed (WDM) system in which a number of optical channels are carried over a common path at disparate wavelengths. It will be understood that the optical communication system 10 may comprise other suitable single channel, multichannel or bi-directional transmission systems.

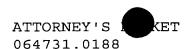
Referring to FIGURE 1, the WDM system 10 includes a WDM transmitter 12 at a source end point and a WDM receiver 14 at a destination end point coupled together The WDM transmitter 12 transmits by an optical link 16. data in a plurality of optical signals, or channels, over the optical link 16 to the remotely located WDM receiver Spacing between the channels is selected to avoid or minimize cross talk between adjacent channels. embodiment, as described in more detail below, minimum spacing (df) comprises a multiple channel transmission symbol and/or bit rate (B) within 0.4 to 0.6 (N). Expressed mathematically: of an integer This neighboring (N+0.4) B < df < (N+0.6) B. suppresses It will be understood that channel channel cross talk. spacing may be suitably varied without departing from the scope of the present invention.

WDM transmitter 12 includes a plurality 20 and a WDM multiplexer 22. optical senders optical sender 20 generates an optical information signal 24 on one of a set of distinct wavelengths  $\lambda_1$ ,  $\lambda_2$  ...  $\lambda_n$ at the channel spacing. The optical information signals signals with least one 24 comprise optical at characteristic modulated to encode audio, video, textual,

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real-time, non-real-time or other suitable data. The optical information signals 24 are multiplexed into a single WDM signal 26 by the WDM multiplexer 22 for transmission on the optical link 16. It will be understood that the optical information signals 24 may be otherwise suitably combined into the WDM signal 26. The WDM signal is transmitted in the synchronous optical network (SONET) or other suitable format.

The WDM receiver 14 receives, separates and decodes the optical information signals 24 to recover In one embodiment, the WDM receiver 14 included data. includes a WDM demultiplexer 30 and a plurality demultiplexer optical receivers 32. The MDMdemultiplexes the optical information signals 24 from the single WDM signal 26 and sends each optical information signal 24 to a corresponding optical receiver 32. optical receiver 32 optically or electrically recovers the encoded data from the corresponding signal 24. used herein, the term each means every one of at least a subset of the identified items.

The optical link 16 comprises optical fiber or other suitable medium in which optical signals may be transmitted with low loss. Interposed along the optical link 16 are one or more optical amplifiers 40. The optical amplifiers 40 increase the strength, or boost, one or more of the optical information signals 24, and thus the WDM signal 26, without the need for optical-to-electrical conversion.

optical amplifiers 40 embodiment, the In one and distributed 30 comprise discrete amplifiers 42 amplifiers 44. The discrete amplifiers 42 comprise rare earth doped fiber amplifiers, such as erbium doped fiber amplifiers (EDFAs), and other suitable amplifiers

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operable to amplify the WDM signal 26 at a point in the optical link 16.

The distributed amplifiers 44 amplify the WDM signal 26 along an extended length of the optical link 16. In one embodiment, the distributed amplifiers 44 comprise bi-directional distributed Raman amplifiers (DRA). Each bi-directional DRA 44 includes a forward, or co-pumping source laser 50 coupled to the optical link 16 at a beginning of the amplifier 44 and a backward, or counterpumping source laser 52 coupled to the optical link 16 at an end of the amplifier 44. It will be understood that the co-pumping and counter-pumping source lasers 50 and 52 may amplify disparate or only partially overlapping lengths of the optical link 16.

50 and 52 comprise sources Raman pump other suitable lasers capable semiconductor orgenerating a pump light, or amplification signal, capable of amplifying the WDM signal 26 including one, more or all of the included optical information signals 24. pump sources 50 and 52 may be depolarized, polarization multiplexed to scrambled polarization minimize orpolarization sensitivity of Raman gain.

The amplification signal from the co-pumping laser 52 is launched in the direction of travel of the WDM signal 26 and thus co-propagated with the WDM signal 26 at substantially the same speed and/or a slight or other suitable velocity mismatch. The amplification signal from the counter-pumping laser 52 is launched in a direction of travel opposite that of the WDM signal 26 and thus is counter-propagated with respect to the WDM signal 26. The amplification signals may travel in opposite directions simultaneously at the same or other suitable speed.

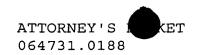
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The amplification signals comprise one or more high power lights or waves at a lower wavelength than the signal or signals to be amplified. As the amplification signal travels in the optical link 16, it scatters off atoms in the link 16, loses some energy to the atoms and continues with the same wavelength as the amplified signal or signals. In this way, the amplified signal acquires energy over many miles or kilometers in that it is represented by more photons. For the WDM signal 26, the co-pumping and counter-pumping lasers 50 and 52 may each comprise several different pump wavelengths that are used together to amplify each of the wavelength distincts optical information signals 24.

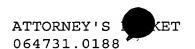
In one embodiment, as described in more detail below, a non-intensity characteristic of a carrier signal is modulated with the data signal at each optical sender The non-intensity characteristic comprises phase, 20. frequency or other suitable characteristic with no or limited susceptibility to cross talk due to cross-gain from a forward pumping distributed modulation (XGM) bi-directional pumping distributed amplifier ora modulated amplifier. The non-intensity information signal may be further and/or remodulated with a clock or other non-data signal using an intensity Thus, the non-intensity modulated optical information signal may comprise intensity modulation of a non-data signal.

In a particular embodiment, as described in more detail below, the WDM signal 26 comprises phase or frequency modulated optical information signals 24 which are amplified using the bi-directional DRAs 44 with no cross talk between the channels 24 due to XGM. In this embodiment, the bi-directional DRAs 44 provide

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amplification at a superior optical signal-to-noise ratio and thus enable longer transmission distances and improved transmission performance.

FIGURE 2 illustrates details of the optical sender 20 in accordance with one embodiment of the present invention. In this embodiment, the optical sender 20 comprises a laser 70, a modulator 72 and a data signal 74. The laser 70 generates a carrier signal at a prescribed frequency with good wavelength control. Typically, the wavelengths emitted by the laser 70 are selected to be within the 1500 nanometer (nm) range, the range at which the minimum signal attenuation occurs for silica-based optical fibers. More particularly, the wavelengths are generally selected to be in the range from 1310 to 1650 nm but may be suitably varied.

The modulator 72 modulates the carrier signal with the data signal 74 to generate the optical information signal 24. The modulator 72 may employ amplitude modulation, frequency modulation, phase modulation, intensity modulation, amplitude-shift keying, frequency-shift keying, phase-shift keying and other suitable techniques for encoding the data signal 74 onto the carrier signal. In addition, it will be understood that different modulators 72 may employ more than one modulation system in combination.

modulates the phrase, frequency or other suitable non-intensity characteristic of the carrier signal with the data signal 74. As previously described, this generates a non-intensity optical information signal 24 with poor susceptibility to cross talk due to XGM in long-haul and other transmission systems using bi-directional DRA or other distributed amplification. Details of the carrier

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wave, frequency modulation of the carrier wave and phase modulation of the carrier wave are illustrated in FIGURES 3A-C.

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Referring to FIGURE 3A, the carrier signal 76 is a completely periodic signal at the specified wavelength. The carrier signal 76 has at least one characteristic that may be varied by modulation and is capable of carrying information via modulation.

Referring to FIGURE 3B, the frequency of the carrier signal 76 is modulated with a data signal 74 to generate a frequency modulated optical information signal 78. In frequency modulation, the frequency of the carrier signal 76 is shifted as a function of the data signal 74. Frequency shift keying may be used in which the frequency of the carrier signal shifts between discrete states.

Referring to FIGURE 3C, the phase of the carrier signal 76 is modulated with a data signal 80 to generate a phase modulated optical information signal 82. In phase modulation, the phase of the carrier signal 76 is shifted as a function of the data signal 80. Phase shift keying may be used in which the phase of the carrier signal shifts between discrete states.

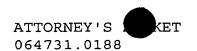
an optical sender 80 in FIGURE illustrates another embodiment of the accordance with this embodiment, data is phase invention. In frequency modulated onto the carrier signal and then remodulated with intensity modulation synchronized with the signal clock to provide superior power tolerance in the transmission system.

Referring to FIGURE 4, the optical sender 80 includes a laser 82, a non-intensity modulator 84 and data signal 86. The non-intensity modulator 84 modulates the phase or frequency of the carrier signal from the

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laser 82 with the data signal 86. The resulting data modulated signal is passed to the intensity modulator 88 for remodulation with the clock frequency 90 to generate a dual or otherwise multimodulated optical information signal 92. Because the intensity modulation based on the clock is a non-random, completely periodic pattern, little or no cross talk due to XGM is generated by the DRAs 44 so long as there is a slight velocity mismatch in the forward pumping direction. FIGURE 5 illustrates the waveform of the dual modulated optical information signal 92.

FIGURE 6 illustrates details of the optical receiver 32 in accordance with one embodiment of the present invention. In this embodiment, the optical receiver 32 receives a demultiplexed optical information signal 24 with the data modulated on the phase of the carrier signal with phase shift keying. It will be understood that the optical receiver 32 may be otherwise suitably configured to receive and detect data otherwise encoded in an optical information signal 24 without departing from the scope of the present invention.

Referring to FIGURE 6, the optical receiver includes an asymmetric interferometer 100 and a detector The interferometer 100 is an asymmetric Mach-Zender or other suitable interferometer operable to convert a non-intensity modulated optical information signal into an intensity modulated optical information signal for detection of data by the detector 102. Preferably, interferometer 100 with Mach-Zender wavelength the dependent loss and good rejection characteristics for the channel spacing.

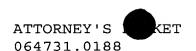
The Mach-Zender interferometer 100 splits the received optical signal into two interferometer paths 110

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and 112 of different lengths and then combines the two paths 110 and 112 interferometrically to generate two complimentary output signals 114 and 116. In particular, the optical path difference (L) is equal to the symbol rate (B) multiplied by the speed of light (c) and divided by the optical index of the paths (n). Expressed mathematically: L=Bc/n.

In a particular embodiment, the two path lengths 110 and 112 are sized based on the symbol, or bit rate to provide a one symbol period, or bit shift. In this embodiment, the Mach-Zender interferometer 100 has a wavelength dependent loss that increases the rejection of neighboring channels when channel spacing comprises the symbol transmission rate multiple within 0.4 to 0.6 of an integer as previously described.

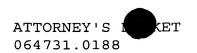
The detector 102 is a dual or other suitable In one embodiment, the dual detector 102 detector. includes photodiodes 120 and 122 connected in series in a balanced configuration and a limiting amplifier 124. this embodiment, the two complimentary optical outputs 114 and 116 from the Mach-Zender interferometer 100 are applied to the photodiodes 120 and 122 for conversion of the optical signal to an electrical signal. The limiting electronic amplifier 124 converts the electrical signal to a digital signal (0 or 1) depending on the optical the interferometer 100. delivered by another embodiment, the detector 102 is a single detector with one photodiode 122 coupled to output 116. embodiment, output 114 is not utilized.

FIGURE 7 illustrates the frequency response of the asymmetric Mach-Zender interferometer 100 in accordance with one embodiment of the present invention. In this embodiment, channel spacing comprises the symbol

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transmission rate multiple within 0.4 to 0.6 of an integer as previously described. As can be seen, optical frequency of neighboring channels is automatically rejected by the asymmetric Mach-Zender interferometer 100 to aid channel rejection of the demultiplexer 30. It will be understood that the asymmetric Mach-Zender interferometer may be used in connection with other suitable channel spacings.

FIGURES 8A-C illustrate details of the demultiplexer 30 in accordance with one embodiment of the present invention. In this embodiment, phase or frequency modulated optical information signals 24 are converted to intensity modulate optical information signals within the demultiplexer 30 of the WDM receiver 14 and/or before demultiplexing or between demultiplexing steps. It will be understood that the demultiplexer 30 may otherwise suitably demultiplex and/or separate the optical information signals 24 from the WDM signal 26 without departing from the scope of the present invention.

demultiplexer FIGURE 8A, the 20 Referring to comprises a plurality of demultiplex elements 130 and a multi-channel format converter 131. Each demultiplex element 130 separates a received set of channels 132 into two discrete sets of channels 134. Final separation is performed by dielectric filters 136 which 25 each filter a specific channel wavelength 138.

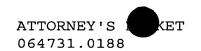
The multichannel format converter 131 converts phase modulation to intensity modulation and may be an asymmetric Mach-Zender interferometer with a one-bit shift to convert non-intensity modulated signals to intensity modulated signals as previously described in connection with interferometer 100 or suitable optical device having a periodical optical frequency response

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that converts at least two phase or frequency modulated channels into intensity modulated WDM signal channels. The intensity-conversion interferometer may be prior to the first stage demultiplex element 130, between the first and second stages or between other suitable stages. The other demultiplex elements 130 may comprise filters or non-conversion Mach-Zender interferometers operable to filter the incoming set of channels 132 into the two sets of output channels 134.

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In a particular embodiment, the multichannel format converter 131 is an asymmetric Mach-Zender interferometer with a free spectral range coinciding with the WDM channel spacing or its integer sub-multiple. all the WDM channels to be converted within the Machsimultaneously. In this interferometer Zender embodiment, a channel spacing may be configured based on the channel bit rate which defines the free spectral Placement of the intensity-conversion Mach-Zender interferometer in the demultiplexer 30 eliminates the need for the interferometer 100 at each optical receiver 32 which can be bulky and expensive. In addition, the demultiplexer 30 including the Mach-Zender and other demultiplexer elements 130 may be fabricated on a same chip which reduces the size and cost of the WDM receiver 14.

8B, the demultiplexer 30 Referring to FIGURE comprises a plurality of wavelength interleavers 133 and a multichannel format converter 135 for each set of by the interleaved optical information signals output last stage wavelength interleavers 133. Each wavelength interleaver 133 separates a received set of channels into interleaved channels. The discrete sets of two multichannel format converters 135 may be asymmetric

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Mach-Zender interferometers with a one-bit shift convert non-intensity modulated signals to intensity modulated signals as previously described in connection with interferometer 100 or other suitable optical device. Use of the wavelength interleavers as part of the WDM demultiplexing in front of the format converters allow several WDM channels to be converted simultaneously in one Mach-Zender interferometer even if the free spectral range of the interferometer does not coincide with an integer multiple of the WDM channel spacing. FIGURE 8C four of Mach-Zender transmissions illustrates of for a particular embodiment interferometers demultiplexer 30 using wavelength interleavers 133 which the free spectral range is three quarters of the In this embodiment, the four Machchannel spacing. Zender interferometers may be used to convert all of the WDM channels.

FIGURE 9 illustrates a method for transmitting information in an optical communication system using distributed amplification in accordance with one embodiment of the present invention. In this embodiment, data signals are phase-shift keyed onto the carrier signal and the signal is amplified during transmission using discrete and distributed amplification.

Referring to FIGURE 9, the method begins at step 140 in which the phase of each disparate wavelength optical carrier signal is modulated with a data signal 74 to generate the optical information signals 24. At step 142, the optical information signals 24 are multiplexed into the WDM signal 26. At step 143, the WDM signal 26 is transmitted in the optical link 16.

Proceeding to step 144, the WDM signal 26 is amplified along the optical link 16 utilizing discrete

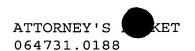
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and distributed amplification. As previously described, the WDM signal 26 may amplified at discrete points using EDFAs 42 and distributively amplified using bidirectional DRAs 44. Because the data signals are modulated onto the phase of the carrier signal, cross talk between channels from XGM due to forward pumping amplification is eliminated. Accordingly, the signal-to-noise ratio can be maximized and the signals may be transmitted over longer distances without regeneration.

Next, at step 145, the WDM signal 26 is received by the WDM receiver 14. At step 146, the WDM signal 26 is demultiplexed by the demultiplexer 30 to separate out the optical information signals 24. At step 147, the phase modulated optical information signals 24 are converted to intensity modulated signals for recovery of the data signal 74 at step 148. In this way, data signals 74 are transmitted over long distances using forward or bidirectional pumping distributed amplification with a low bit-to-noise ratio.

illustrates a bi-directional optical FIGURE 10 system 150 in accordance with communication embodiment of the present invention. In this embodiment, the bi-directional communication system 150 includes WDM transmitters 152 and WDM receivers 154 at each end of an The WDM transmitters 152 comprise optical link 156. optical senders and a multiplexer as previously described in connection with the WDM transmitter 12. the WDM receivers 154 comprise demultiplexers and optical receivers as previously described in connection with the WDM receiver 14.

At each end point, the WDM transmitter and receiver set is connected to the optical link 156 by a routing device 158. The routing device 158 may be an optical

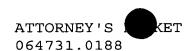
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circulator, optical filter, or optical interleaver filter capable of allowing egress traffic to pass onto the link 156 from WDM transmitter 152 and to route ingress traffic from the link 156 to WDM receiver 154.

156 comprises bi-directional link The optical discrete amplifiers 160 and bi-directional distributed amplifiers 162 spaced periodically along the link. bi-directional discrete amplifiers 160 may comprise EDFA amplifiers as previously described in connection with amplifiers 42. Similarly, the distributed amplifiers 162 may comprise DRA amplifiers including co-pumping 166 previously 164 and as counter-pumping lasers described in connection with DRA amplifiers 44.

WDM signal is generated operation, a transmitted from each end point to the other end point and a WDM signal is received from the other end point. Along the length of the optical link 156, the WDM signals amplified using bi-directional-pumped 162. DRA Because data is not carried in the form of optical intensity, cross talk due to XGM is eliminated. DRA and other suitable distributed amplification may be used in long-haul and other suitable bi-directional optical transmission systems.

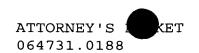
FIGURE 11 illustrates an optical sender 200 and an accordance 202 in with receiver optical embodiment of the present invention. In this embodiment, the optical sender 200 and the optical receiver fine-tune modulation for communicate to optical information transmission performance of the It will be understood that modulation of the signals 24. optical information signals 24 may be otherwise finetuned using downstream feedback without departing from the scope of the present invention.

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Referring to FIGURE 11, the optical sender comprises a laser 210, a modulator 212, and a data signal 214 which operate as previously described in connection with the laser 70, the modulator 72 and the data signal A controller 216 receives bit error rate or other indication of transmission errors from the downstream optical receiver 202 and adjust the modulation depth of modulator 212 based on the indication to reduce and/or minimize transmission errors. The controller 216 may adjust the amplitude, intensity, phase, frequency and/or other suitable modulation depth of modulator 212 and may use any suitable control loop or other algorithm that adjusts modulation alone or in connection with other characteristics minimized toward а or transmission error rate. Thus, for example, may adjust a non-intensity modulation controller 216 depth and a depth of the periodic intensity modulation in sender 80 generate and optimize optical to the multimodulated signals.

The optical receiver 202 comprises an interferometer 220 and a detector 222 which operate as previously described in connection with interferometer 100 and detector 102. A forward error correction (FEC) decoder 224 uses header, redundant, symptom or other suitable bits in the header or other section of a SONET or other frame or other transmission protocol data to determine bit errors. The FEC decoder 224 corrects for detected bit errors and forwards the bit error rate or other indicator of transmission errors to a controller 226 for the optical receiver 202.

The controller 226 communicates the bit error rate or other indicator to the controller 216 in the optical sender 200 over an optical supervisory channel (OSC) 230.

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The controllers 216 and 226 may communicate with each other to fine-tune modulation depth during initiation or setup of the transmission system, periodically during operation of the transmission system, continuously during operation of the transmission system or in response to predefined trigger events. In this way, modulation depth is adjusted based on received signal quality measured at the receiver to minimize chromatic dispersion, non-linear effects, receiver characteristics and other unpredictable and/or predictable characteristics of the system.

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FIGURE 12 illustrates details of the modulator 212 in accordance with one embodiment of the present invention. In this embodiment, the modulator 212 employs phase and intensity modulation to generate a bi-modulated optical information signal. The phase and intensity modulation depth is adjusted based on receiver-side feedback to minimize transmission errors.

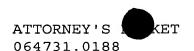
Referring to FIGURE 12, the modulator 212 includes for phase modulation such as phase shift keying a bias circuit 230 coupled to an electrical driver 232. bias circuit 230 may be a power supply and the electrical driver 232 a broadband amplifier. The bias circuit 230 is controlled by the controller 216 to output a bias signal to the electrical driver 232. The bias signal provides an index for phase modulation. The electrical driver 232 amplifies the data signal 214 based on the bias signal and outputs the resulting signal to phase modulator 234. Phase modulator 234 modulates the receive bias-adjusted data signal onto the phase of the carrier signal output by the laser 210 to generate a phase modulated optical information signal 236.

For intensity modulation such as intensity shift keying, the modulator 212 includes a bias circuit 240

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coupled to an electrical driver 242. The bias circuit 240 is controlled by the controller 216 to output a bias signal to the electrical driver 242. The bias signal acts as an intensity modulation index. The electrical driver 242 amplifies a network, system or other suitable clock signal 244 based on the bias signal and outputs the resulting signal to the intensity modulator 246. intensity modulator 246 is coupled to the phase modulator 234 and modulates the receive bias-adjusted clock signal onto the phase modulated optical information signal 236 to generate the bi-modulated optical information signal It will be understood for transmission to a receiver. that phase and intensity modulation at the transmitter may be otherwise suitably controlled based on receiverside feedback to minimize transmission errors of data over the optical link.

FIGURE 13 illustrates a method for fine tuning modulation depth of an optical information signal using accordance information in with one side receiver embodiment of the present invention. The method begins at step 250 in which an optical carrier is modulated with a data signal 214 at the optical sender 200. Next, at step 252, the resulting optical information signal 24 is transmitted to the optical receiver 202 in a WDM signal 26.

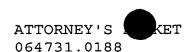
Proceeding to step 254, the data signal 214 is recovered at the optical receiver 204. At step 256, the FEC decoder 224 determines a bit error rate for the data based on bits in the SONET overhead. At step 258, the bit error rate is reported by the controller 226 of the optical receiver 202 to the controller 216 of the optical sender 200 over the OSC 230.

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Next, at decisional step 260, the controller 216 determines whether modulation is optimized. In one embodiment, modulation is optimized when the bit error rate is minimized. If the modulation is not optimized, the No branch of decisional step 260 leads to step 262 in which the modulation depth is adjusted. Step 262 returns to step 250 in which the data signal 214 is modulated with the new modulation depth and transmitted to the optical receiver 202. After the modulation depth is optimized from repetitive trails and measurements or other suitable mechanisms, the Yes branch of decisional step 260 leads to the end of the process. In this way, transmission performance is improved and transmission errors minimized.

FIGURE 14 illustrates an optical communication system 275 distributing a clock signal in an information channel in accordance with one embodiment of the present invention. In this embodiment, pure clock is transmitted in channels to one, more or all nodes in the optical system 275.

Referring to FIGURE 14, optical system 275 includes a WDM transmitter 280 coupled to a WDM receiver 282 over an optical link 284. The WDM transmitter 280 includes a plurality of optical senders 290 and a WDM multiplexer Each optical sender 290 generates an optical on one of a set of discrete information signal 294 wavelengths at the channel spacing. In the clock channel sender 290 generates the optical information signal 294 with at least one characteristic modulated to encode the clock signal. In the data channels 297, the optical sender 290 generates an optical information signal 294 with at least one characteristic modulated to encode a corresponding data signal.

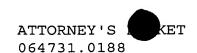
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The optical signals 294 from the clock and data channels 296 and 297 are multiplexed into a signal WDM signal 298 by the WDM multiplexer 292 for transmission on the optical link 284. Along the optical link 284, the signal may be amplified by discrete and/or distributed amplifiers as previously described.

The WDM receiver 282 receives, separates and decodes the optical information signals 294 to recover the included data and clock signals. In one embodiment, the WDM receiver 282 includes a WDM demultiplexer 310 and a plurality of optical receivers 312. The WDM demultiplexer 310 demultiplexes the optical information signals 294 from the single WDM signal 298 and sends each optical information signal 294 to a corresponding optical receiver 312.

Each optical receiver 312 optically or electrically recovers the encoded data or clock signal from the corresponding signal 294. In the clock channel 296, the clock signal is recovered and forwarded to the optical receivers 312 in the data channels 297 for use in data The extraction and forward error correction. transmission of pure clock in an information allows a more stable clock recovery with less jitter. The stable clock may be used by forward error correction to improve the bit error rate even in the presence of jitter and poor optical signal quality.

FIGURE 15 illustrates an optical receiver 320 for extracting a clock signal from a multimodulated signal in accordance with one embodiment of the present invention. In this embodiment, the optical receiver 320 receives a demultiplexed optical information signal with data phase modulated onto a carrier signal that is then remodulated with intensity modulation synchronized with the network,

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system or other suitable clock as described in connection with the optical sender 80. The optical receiver 320 extracts the clock information from the optical signal and uses the stable clock to recover data from the phase modulated signal of the channel. Thus, each channel can recover its own clock.

Referring to FIGURE 15, the optical receiver 320 includes an interferometer 322 and a detector 324 as previously described in connection with the optical receiver 32. The interferometer 322 receives the miltimodulated signal and converts the phase modulation into intensity modulation for recovery of the data signal 330 by the detector 324.

A clock recovery element 326 comprises a photodiode and/or other suitable components to recover the clock signal before phase-to-intensity conversion of the data signal. The clock recovery element 326 may comprise a phase lock loop, a tank circuit, a high quality filter and the like. The clock recovery element 326 receives the multimodulated signal and recovers the clock signal 332 from the intensity modulation.

The data signal 330 and the recovered clock signal 332 are output to a digital flip flop or other suitable data recovery circuit 334. In this way, the optical receiver 320 extracts the clock information from the optical signal before the phase-to-intensity conversion of the data signal and provides a stable clock recovery with less jitter even with poor optical signal quality corresponding to a bit error rate in the range of 1e<sup>-2</sup>.

Although the present invention has been described with several embodiments, various changes and modifications may be suggested to one skilled in the art. It is intended that the present invention encompass such

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changes and modifications as fall within the scope of the appended claims.